Design of a Wide Upstream-Bandwidth Optical Triplexer Using Three-Waveguide Interferometer

Jun-Seok Choi and Hyung-Jong Lee Chonnam National University 300 Yong-Bong Dong Buk-Gu Gwang-Ju 500-757 Republic of Korea Email: etrx@naver.com

Abstract—A novel planar lightwave circuit (PLC) triplexer is designed using a three-waveguide interferometer and Mach-Zehnder interferometers. We introduce a full-coupling scheme in the three-waveguide coupler for wide upstream-bandwidth characteristic. Transfer matrix forms for three-waveguide coupler and three-waveguide phase shifter are derived for numerical simulation. Simulation results show that the 1dB- and 3dBbandwidth for the upstream channel are 115nm and 200nm, respectively. The proposed triplexer shows low insertion losses of less than 0.6dB and low crosstalks of less than -28dB for the two downstream bands, assuming that the waveguide width-error in the PLC fabrication is $\pm 0.1 \mu m$.

I. INTRODUCTION

Bi-directional optical triplexer is one of key components in the passive optical network (PON) system for FTTH (fiberto-the-home) application, providing extra services like video or others in addition to the common service of data. It is required for the triplexer to demultiplex the two closelyspaced wavelength-bands of 1550±10nm and 1490±10nm in downstream and to multiplex the one far-off wavelength-band of 1310±50nm in upstream according to the standard of the International Telecommunication Union(ITU) T-983.3.

Recently, several studies have been made on planar lightwave circuit (PLC) triplexer [1], [2]. However, these have difficulties in satisfying the bandwidth requirement for the upstream channel. A triplexer composed of series of Mach-Zehnder interferometers (MZIs) is also proposed [3]. This type of triplexer must use at least two stages of MZIs for separating the three channels, because the MZI is only a two-port device in the input or output.

In this work, we propose a three-waveguide interferometer (TWI) for optical triplexer application. Trasfer matrix forms for three-waveguide coupler and three-waveguide phase shifter are derived. We introduce a full-coupling scheme for the far-off upstream channel (1310 nm) in the three-waveguide coupler of the TWI, and were able to design a compact low-loss PLC triplexer which has wide bandwidth in the upstream channel.

II. DESIGN AND SIMULATION

The proposed trixer consists of three parts as shown in Fig. 1: A three-waveguide interferometer (TWI), a demultiplexing Seo-Young Lee

Gwang-Ju Institute of Science and Technology 261 Cheomdan-Gwagiro Buk-Gu Gwang-Ju 500-712 Republic of Korea

Email: zz123z@daum.net

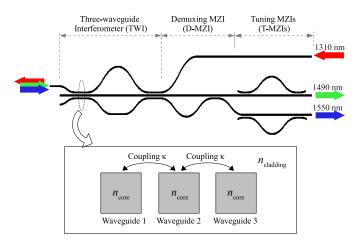


Fig. 1. Overall design of the triplexer and cross-sectional view of the threewaveguide coupler.

Mach-Zehnder interferometer (D-MZI), and tuning MZIs (T-MZIs).

The TWI is designed to demultiplex the one far-off upstream channel (1310nm) and the two closely-spaced downstream channels (1490- and 1550-nm). From the coupled-mode equation, the transfer matrices of the three-waveguide coupler and phase shifter can be derived as follows

$$[M_{coupler}] = \begin{pmatrix} b + 0.5 & a & b - 0.5 \\ a & 2b & a \\ b - 0.5 & a & b + 0.5 \end{pmatrix}$$
(1)

$$[M_{coupler}] = \begin{pmatrix} b+0.5 & a & b-0.5 \\ a & 2b & a \\ b-0.5 & a & b+0.5 \end{pmatrix}$$
(1)
$$[M_{ps}] = \begin{pmatrix} \exp(j\beta L_1) & 0 & 0 \\ 0 & \exp(j\beta L_2) & 0 \\ 0 & 0 & \exp(j\beta L_3) \end{pmatrix}$$
(2)

where, $a=-\frac{j}{\sqrt{2}}\sin(\sqrt{2}\kappa L_{cpl}),~b=\frac{1}{2}\cos(\sqrt{2}\kappa L_{cpl}),$ and L_{cpl} and L_i denotes the length of the coupler and the length of the ith waveguide in phase shifter. κ and β are the coupling constant and the propagation constant of the waveguide.

The transfer matrix form of the three-waveguide interferometer is then given by

$$[M_{TWI}] = [M_{coupler}] [M_{ps}] [M_{coupler}]$$
(3)

In the TWI operation, the bandwidth for the upstream signal (λ_1 =1310nm) must be as wide as 100nm while the band separation of 60nm between two downstream signals (λ_2 =1490nm and λ_3 =1550nm). However, when the TWI separates each wavelength by interference, the bandwidth for the upstream becomes narrower.

To solve this problem, we set the coupling length L_{cpl} of the three-waveguide coupler to be integer times of the full-coupling length, which is expressed by

$$L_{cpl} = \frac{m\pi}{\sqrt{2}\kappa(\lambda_1)}, \ (m = 1, 2, 3, \cdots)$$
 (4)

On the other hands, the λ_2 and λ_3 signals are distributed into the three waveguides of the phase shifter and interfere in the second coupler. There is a simple analytic solution for λ_3 signal to go to the output port 3, which is given by

$$\kappa(\lambda_3)L_{cpl} = (m + \frac{1}{2})\frac{\pi}{\sqrt{2}} \tag{5}$$

$$\cos(\beta \Delta L_{12}) = \cos(\beta \Delta L_{23}) = \cos(\beta \Delta L_{31}) = 1$$
 (6)

where, ΔL_{ij} denotes the path length difference between the *i*th and *j*th waveguides in the phase shifter.

However, there is no analytic solution for λ_2 to be 1 in output port 2. Therefore, we numerically searched for the optimum parameters such as κ , L_{cpl} , ΔL_{12} , and ΔL_{23} for the separation of λ_2 and λ_3 into the output port 2 and 3, respectively.

In this work, ΔL_{12} and ΔL_{23} are determined to be $26.69\mu m$ and $-12.85\mu m$, respectively. The coupling length L_{cpl} and the distance between adjacent waveguides in the coupler are determined to be $664\mu m$ and $1\mu m$, respectively, from Eq. (4) and Eq. (5).

An additional demultiplexing device is combined to the output port 2 and 3 of the TWI to improve the insertion loss characteristic of downstream channels as shown in Fig. 1. Here, the device is a single-stage MZI, called demuxing MZI (D-MZI). We numerically searched for the optimized design parameters such as coupling length and path-length difference for the D-MZI to have the maximum at the wavelength of λ_2 and λ_3 in the output port 2 and 3, respectively. Also, tuning MZIs (T-MZIs) are concatenated after the D-MZI for further enhancement of the performances such as fabrication tolerance and polarization dependence. They consist of two directional couplers and a phase shifter. The T-MZIs at the output port 2 and port 3 are designed to have constructive interference for λ_2 and λ_3 , respectively, and destructive interference for λ_3 and λ_2 , respectively.

The spectral response of the proposed triplexer simulated by the transfer matrix method is shown in Fig. 2. The crosstalk dependence and the insertion-loss dependence on polarization and width-error of the waveguide are shown in Fig. 3. Assuming the waveguide width-error is less than $\pm 0.1 \mu m$, the insertion losses and the crosstalks of the proposed triplexer are maintained to be less than 0.6dB and -28dB at the center wavelength of each wavelength band. The tolerance $\pm 0.1 \mu m$ of the waveguide width-error can be commonly achieved using standard lithographic processes in silicon electronics industry.

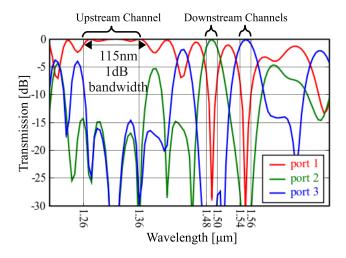


Fig. 2. Spectral response of the proposed triplexer.

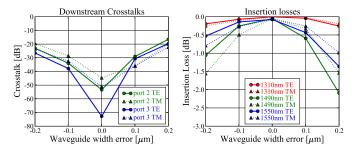


Fig. 3. Effects of polarization and waveguide width-error of the triplexer.

III. CONCLUSION

We proposed a PLC-type triplexer which consists of TWI and MZIs. Simulation and optimization for the proposed triplexer are performed by using the transfer matrix method. For wide-bandwidth operation for the upstream channel, full-coupling scheme was introduced. Simulation results show that 1dB- and 3dB-bandwidth of the upstream channel are 115nm and 200nm, respectively.

To improve the performances of the triplexer, we combined a demultiplexing MZI and tuning MZIs to the downstream ports of the TWI. The insertion losses and crosstalks are maintained to be less than 0.6dB and -28dB, respectively, at the center wavelength of each wavelength band, assuming the waveguide width-error of $\pm 0.1 \mu m$.

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